| DESCRIP turing syn ture allow input, the a variety modes, st | ION hrono the ' 1 ermin meth cha | 54/74190 <br> 54LS/74LS <br> /DOWN DECADE <br> (With Preset and Rip <br> The '190 is a reversible B us counting and asynchron 0 to be used in programm Count output and the Rip ds of implementing multiges are initiated by the ris | ple Clock) <br> CD (8421) decade counter nous presetting. The pres able dividers. The Count E ple Clock output make po stage counters. In the cou ising edge of the clock. |  | CONNECTION DIAGRAM PINOUT A |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DESCRIPTION - The '190 is a reversible BCD (8421) decade turing synchronous counting and asynchronous presetting. T ture allows the '190 to be used in programmable dividers. The input, the Terminal Count output and the Ripple Clock output a variety of methods of implementing multi-stage counters. I modes, state changes are initiated by the rising edge of the <br> - HIGH SPEED - 30 MHz TYPICAL COUNT FREQUENCY <br> - SYNCHRONOUS COUNTING <br> - ASYNCHRONOUS PARALLEL LOAD <br> - CASCADABLE |  |  |  |  | LOGIC SYMBOL |
|  |  | COMMERCIAL GRADE | MILITARY GRADE |  |  |
| PKGS |  | $\begin{aligned} & V_{C C}=+5.0 \mathrm{~V} \pm 5 \%, \\ & T_{\mathrm{A}}=0^{\circ} \mathrm{C} \text { to }+70^{\circ} \mathrm{C} \end{aligned}$ | $\begin{gathered} V_{C C}=+5.0 \mathrm{~V} \pm 10 \% \\ T_{A}=-55^{\circ} \mathrm{C} \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ |  | $14-\left[\begin{array}{cccc} c p & & & \\ a_{0} & a_{1} & a_{2} & a_{3} \end{array}\right]{ }^{12}$ |
| Plastic DIP (P) | A | 74190PC, 74LS190PC |  | 9B | $\prod_{3} \prod_{2} \prod_{6} \prod_{7}$ |
| Ceramic DIP (D) | A | 74190DC, 74LS190DC | 54190DM, 54LS190DM | 7B | $\begin{aligned} & \mathrm{VCC}=\operatorname{Pin} 16 \\ & \mathrm{GND}=\operatorname{Pin} 8 \end{aligned}$ |
| Flatpak (F) | A | 74190FC, 74LS190FC | 54190FM, 54LS190FM | 4L |  |

INPUT LOADING/FAN-OUT: See Section 3 for U.L. definitions

| PIN NAMES | DESCRIPTION | 54/74 (U.L.) <br> HIGH/LOW | 54/74LS (U.L.) <br> HIGH/LOW |
| :---: | :---: | :---: | :---: |
| $\overline{\mathrm{CE}}$ | Count Enable Input (Active LOW) | 3.0/3.0 | 1.5/0.75 |
| CP | Clock Pulse Input (Active Rising Edge) | 1.0/1.0 | 0.5/0.25 |
| $\mathrm{P}_{0}-\mathrm{P}_{3}$ | Parallel Data Inputs | 1.0/1.0 | 0.5/0.25 |
| $\overline{\text { PL }}$ | Asynchronous Parallel Load Input (Active LOW) | 1.0/1.0 | 0.5/0.25 |
| $\bar{U} / \mathrm{D}$ | Up/Down Count Control Input | 1.0/1.0 | 0.5/0.25 |
| $\mathrm{Q}_{0}-\mathrm{Q}_{3}$ | Flip-flop Outputs | 20/10 | $\begin{array}{r} 10 / 5.0 \\ (2.5) \end{array}$ |
| $\overline{\mathrm{RC}}$ | Ripple Clock Output (Active LOW) | 20/10 | $\begin{array}{r} 10 / 5.0 \\ (2.5) \end{array}$ |
| TC | Terminal Count Output (Active HIGH) | 20/10 | $\begin{array}{r} 10 / 5.0 \\ (2.5) \end{array}$ |

LOGIC DIAGRAM


MODE SELECT TABLE

| INPUTS |  |  |  | MODE |
| :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { PL }}$ | $\overline{C E}$ | U/D |  |  |
| H | L | L | 」 | Count Up |
| H | L | H | $\checkmark$ | Count Down |
| L | X | X | X | Preset (Asyn.) |
| H | H | X | X | No Change (Hold) |

RC TRUTH TABLE

| INPUTS |  |  | OUTPUT |
| :---: | :---: | :---: | :---: |
| $\overline{\mathrm{CE}}$ | TC* | CP | $\overline{\mathrm{RC}}$ |
| L | H | U | U |
| H | X | X | H |
| X | L | X | H |

H = HIGH Vohage Level
L = LOW Voltage Level
$X=$ Immaterial

STATE DIAGRAM


FUNCTIONAL DESCRIPTION - The '190 is a synchronous up/down BCD decade counter and the '191 is a synchronous up/down 4-bit binary counter. The operating modes of the ' 190 decade counter and the '191 binary counter are identical, with the only difference being the count sequences as noted in the state diagrams. Each circuit contains four master/slave flip-flops, with internal gating and steering logic to provide individual preset, count-up and count-down operations.

Each circuit has an asynchronous parallel load capability permitting the counter to be preset to any desired number. When the Parallel Load ( $\overline{P L}$ input is LOW, information present on the Parallel Data inputs ( $P_{0}-P_{3}$ ) is loaded into the counter and appears on the Q outputs. This operation overrides the counting functions, as indicated in the Mode Select Table.

A HIGH signal on the $\overline{C E}$ input inhibits counting. When $\overline{C E}$ is LOW, internal state changes are initiated synchronously by the LOW-to-HIGH transition of the clock input. The direction of counting is determined by the $\bar{U} / D$ input signal, as indicated in the Mode Select Table. When counting is to be enabled, the $\overline{\mathrm{CE}}$ signal can be made LOW when the clock is in either state. However, when counting is to be inhibited, the LOW-to-HIGH $\overline{C E}$ transition must occur only while the clock is HIGH. Similarly, the $\bar{U} / D$ signal should only be changed when either $\overline{\mathrm{CE}}$ or the clock is HIGH. These restrictions do not apply to the 'LS190 and 'LS191; $\overline{\mathrm{CE}}$ and $\overline{\mathrm{U}} / \mathrm{D}$ can be changed with the clock in either state, provided only that the recommended setup and hold times are observed.

Two types of outputs are provided as overflow/underflow indicators. The Terminal Count (TC) output is normally LOW and goes HIGH when a circuit reaches zero in the count-down mode or reaches maximum ( 9 for the'190, 15 for the '191) in the count-up mode. The TC output will then remain HIGH until a state change occurs, whether by counting or presetting or until $\overline{\mathrm{U}} / \mathrm{D}$ is changed. The TC output should not be used as a clock signal because it is subject to decoding spikes.

The TC signal is also used internally to enable the Ripple Clock ( $\overline{\mathrm{RC}}$ ) output. The $\overline{\mathrm{RC}}$ output is normally HIGH. When $\overline{C E}$ is LOW and TC is HIGH, the $\overline{R C}$ output will go LOW when the clock next goes LOW and will stay LOW until the clock goes HIGH again. This feature simplifies the design of multi-stage counters, as indicated in Figures a and b. In Figure a, each $\overline{\mathrm{RC}}$ output is used as the clock input for the next higher stage. This configuration is particularly advantageous when the clock source has a limited drive capability, since it drives only the first stage. To prevent counting in all stages it is only necessary to inhibit the first stage, since a HIGH signal on $\overline{C E}$ inhibits the $\overline{\mathrm{RC}}$ output pulse, as indicated in the $\overline{\mathrm{RC}}$ Truth Table. A disadvantage of this configuration, in some applications, is the timing skew between state changes in the first and last stages. This represents the cumulative delay of the clock as it ripples through the preceding stages.

A method of causing state changes to occur simultaneously in all stages in shown in Figure b. All clock inputs are driven in parallel and the $\overline{\mathrm{RC}}$ outputs propagate the carry/borrow signals in ripple fashion. In this configuration the LOW state duration of the clock must be long enough to allow the negative-going edge of the carry/borrow signal to ripple through to the last stage before the clock goes HIGH. There is no such restriction on the HIGH state duration of the clock, since the $\overline{\text { RC output of any package goes HIGH shortly after its CP input }}$ goes HIGH.

The configuration shown in Figure $c$ avoids ripple delays and their associated restrictions. The $\overline{\mathrm{CE}}$ input for a given stage is formed by combining the TC signals from all the preceding stages. Note that in order to inhibit counting an enable signal must be included in each carry gate. The simple inhibit scheme of Figures $a$ and $b$ doesn't apply, because the TC output of a given stage is not affected by its own $\overline{\mathrm{CE}}$.


Fig. a N-Stage Counter Using Ripple Clock


Fig. b Synchronous N-Stage Counter Using Ripple Carry/Borrow


Fig. c Synchronous N-Stage Counter with Parallel Gated Carry/Borrow


Fig. d


