SGS-THOMSON MICROELECTRONICS

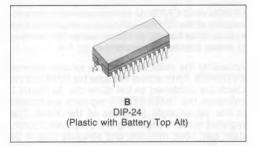
2K × 8 ZEROPOWER™/TIMEKEEPER™ RAM

- INTEGRATED ULTRA LOW POWER SRAM, REAL TIME CLOCK, CRYSTAL, POWER-FAIL CONTROL CIRCUIT AND BATTERY
- BYTEWIDE[™] RAM-LIKE CLOCK ACCESS
- BCD CODED YEAR, MONTH, DAY, DATE, HOURS, MINUTES AND SECONDS
- SOFTWARE CONTROLLED CLOCK CALIBRA-TION FOR HIGH ACCURACY APPLICATIONS
- PREDICTED WORST CASE BATTERY STORAGE LIFE OF 11 YEARS @ 70 °C
- PIN AND FUNCTION COMPATIBLE WITH JEDEC STANDARD 2K × 8 SRAMS
- AUTOMATIC POWER-FAIL CHIP DESELECT/WRITE PROTECTION
- TWO POWER-FAIL DESELECT TRIP POINTS AVAILABLE MK48T02 4.75V \ge V_{PFD} \ge 4.50V MK48T12 4.50V \ge V_{PFD} \ge 4.20V

Part Number	Access Time	R/W Cycle Time
MK48TX2-12	120 ns	120 ns
MK48TX2-15	150 ns	150 ns
MK48TX2-20	200 ns	200 ns
MK48TX2-25	250 ns	250 ns

TRUTH TABLE (MK48T02/12)

V _{cc}	Ē	Ğ	W	MODE	DQ
<v<sub>CC (Max) >V_{CC} (Min)</v<sub>	V _{IH} V _{IL} V _{IL} V _{IL}	X X V _{IL} V _{IH}	X V _{IL} V _{IH} V _{IH}	Deselect Write Read Read	High-Z D _{IN} D _{OUT} High-Z
<v<sub>PFD (Min) >V_{SO}</v<sub>	X	X	X	Power-Fail Deselect	High-Z
≤V _{SO}	X	X	X	Battery Back-up	High-Z



A,	1 🗆	•	24	V _{cc}
A ₆	2 [A ₈
A _s	3 🗆		22	A ₉
A4	4 🗆		21	W
A ₃	5 🗆		20	G
A2			19	A ₁₀
A,	7 🗆		18	Ē
A ₀	8 🗆		17	DQ,
DQ,	9 🗆		16	DQ ₆
DQ ₁	10 🗆		□ 15	DQ5
DQ2	11 🗆		14	DQ4
GND	12 🗆		□ 13	DQ3

PIN NAMES

A ₀ - A ₁₀	Address Inputs	V _{cc}	+5 V		
Ē	Chip Enable	W	Write Enable		
GND	Ground	G	Output Enable		
DQ ₀ —DQ ₇ Data In/Data Out					

DESCRIPTION

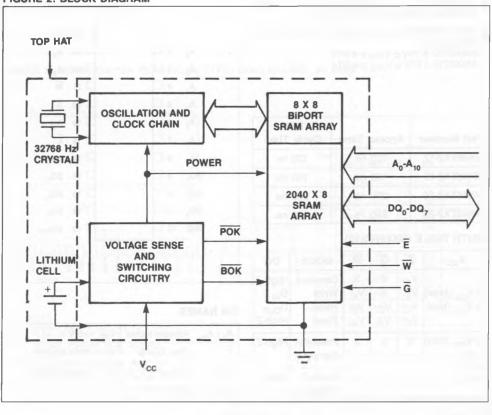
The MK48T02/12 combines a $2K \times 8$ full CMOS SRAM, a BYTEWIDE accessible real time clock, a crystal and a long life lithium carbon mono-fluoride battery, all in a single plastic DIP package. The MK48T02/12 is a non-volatile pin and function equivalent to any JEDEC standard $2K \times 8$ SRAM, such as the 6116 or 5517. It also easily fits into many EPROM AND EEPROM sockets, providing the non-volatility of the PROMs without any requirement for special write timing, or limitations on the number of writes that can be performed.

Access to the clock is as simple as conventional BYTEWIDE RAM access because the RAM and the Clock are combined on the same die. As Figure 2 indicates, the TIMEKEEPER registers are located in the upper eight locations of the RAM. The registers contain, beginning at the top; year, month, date, day, hour, minutes, and seconds data in 24 Hour BCD format. Corrections for 28, 29 (Leap Year), 30 and 31 day months are made automatically. The eighth location is a Control register. These registers are not the actual clock counters; they are BiPORT read/write Static RAM memory locations. The MK48T02/12 includes a clock control circuit that, once every second, dumps the counters into the BiPORT RAM.

Because the Clock Registers are constructed using BiPORT memory cells, access to the rest of the RAM proceeds unhindered by updates to the TIMEKEEPER registers, even if the TIMEKEEPER registers are being updated at the very moment another location in the memory array is accessed.

The MK48T02/12 also has its own Power-fail Detect circuit. The circuit deselects the device whenever V_{CC} is out of range, providing a high degree of data security in the midst of unpredictable system operations brought on by low V_{CC} .

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SGS-THOMSO

FIGURE 2. BLOCK DIAGRAM

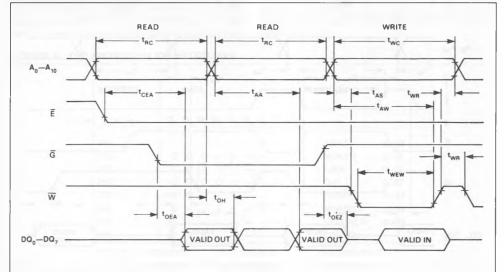
OPERATION

READ MODE

The MK48T02/12 is in the Read Mode whenever \overline{W} (Write Enable) is high and \overline{E} (Chip Enable) is low. The device architecture allows ripple-through access (changing Addresses without removing Chip Enable) to any of the 2048 address locations in the static storage array. Valid data will be available at the Data I/O pins within t_{AA} after the last address input signal is stable, providing that the E and G access times are satisfied.

If \overline{E} or \overline{G} access times are not yet met, valid data will be available at the latter of Chip Enable Access Time (t_{CEA}) or at Output Enable Access Time (t_{OEA}). The state of the eight three-state Data I/O signals is controlled by \overline{E} and \overline{G} . If the Outputs are activated before t_{AA} , the data lines will be driven to an indeterminate state until t_{AA} . If the Address inputs are changed while \overline{E} and \overline{G} remain low, output data will remain valid for Output Data Hold Time (t_{OH}) but will go indeterminant until the next t_{AA} .

FIGURE 3. READ-READ-WRITE TIMING



AC ELECTRICAL CHARACTERISTICS (READ CYCLE TIMING)

 $(0^{\circ}C \le T_A \le 70^{\circ}C) (V_{CC} (Max) \ge V_{CC} \ge V_{CC} (Min))$

		MK48	TX2-12	MK48	TX2-15	MK48	TX2-20	MK48	TX2-25		
SYM	PARAMETER	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	UNITS	NOTES
t _{RC}	Read Cycle Time	120		150		200		250		ns	
t _{AA}	Address Access Time		120		150		200		250	ns	1
tCEA	Chip Enable Access Time		120		150		200		250	ns	1
tOEA	Output Enable Access Time		75		75		80		90	ns	1
t _{CEZ}	Chip Enable Hi to High-Z		30		35		40		50	ns	
t _{OEZ}	Output Enable Hi to High-Z		30		35		40		50	ns	
t _{OH}	Valid Data Out Hold Time	15		15		15		15		ns	1

NOTE

1. Measured using the Output Load Diagram shown in Figure 13.

WRITE MODE

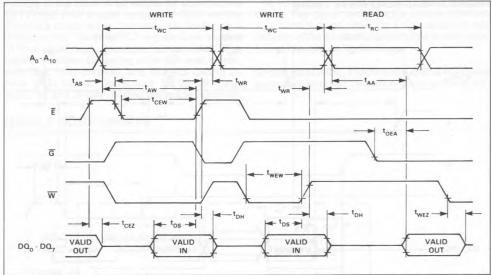
The MK48T02/12 is in Write Mode whenever the W and E inputs are held low. The start of a Write is referenced to the latter occurring falling edge of either Wor E. A Write is terminated by the earlier rising edge of W or E. The addresses must be held valid throughout the cycle. W or E must return high for a minimum of twR prior to the initiation of another Read or Write Cycle. Data-in must be valid for t_{DS} prior to the End of Write and remain valid for t_{DH} afterward.

FIGURE 4. WRITE-WRITE-READ TIMING

Some processors thrash producing spurious Write Cycles during power-up, despite application of a power-on reset. Users should force W or E high during power-up to protect memory after V_{CC} reaches V_{CC} (min) but before the processor stablizes.

The MK48T02/12 G input is a DON'T CARE in the write mode. G can be tied low and two-wire RAM control can be implemented. A low on W will disable the outputs twez after W falls. Take care to avoid bus contention when operating with two-wire control.

NOTES



AC ELECTRICAL CHARACTERISTICS (WRITE CYCLE TIMING)

			MK48TX2-12		MK48TX2-15		MK48TX2-20		MK48TX2-25		
SYM	PARAMETER	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	UNITS	
twc	Write Cycle Time	120		150		200		250		ns	Γ
t _{AS}	Address Setup Time	0		0		0		0		ns	ſ
t _{AW}	Address Valid to End of Write	90		120		140		180		ns	ľ
tCEW	Chip Enable to End of Write	75		90		120		160		ns	ľ
twew	Write Enable to End of Write	75		90		120		160		ns	ľ
t _{WR}	Write Recovery Time	10		10		10		10		ns	Ĩ
t _{DS}	Data Setup Time	35		40		60		100		ns	ľ
t _{DH}	Data Hold Time	0		0		0		0		ns	ſ
twez	Write Enable Low to High-Z		40		50		60		80	ns	ſ

(0°C < T - 70°C) (V. (Max)>V. -11



CLOCK OPERATIONS

Reading the Clock

Updates to the TIMEKEEPER registers should be Halted before clock data is read to prevent reading of data in transition. Because the BiPORT TIMEKEEPER cells in the RAM array are only data registers, and not the actual counters, updating the registers can be halted without disturbing the clock itself.

Updating is halted when a "1" is written into the "Read" bit, the seventh most significant bit in the Control register. As long as a "1" remains in that position, updating is halted. After a Halt is issued, the registers reflect the count, that is day, date, and time that were current at the moment the Halt com-

mand was issued. All of the TIMEKEEPER registers are updated simultaneously. A Halt will not interrupt an update in progress. Updating is within a second after the bit is reset to a "0".

Setting the Clock

The eighth bit of the Control register is the "Write" bit. Setting the Write bit to a "1", like the Read Bit, halts updates to the TIMEKEEPER registers. The user can then load them with the correct day, date and time data in 24 Hour BCD format. Resetting the Write bit to a "0" then transfers those values the actual TIMEKEEPER counters and allows normal operation to resume. The KS bit, FT bit and the bits marked with zeroes in Figure 5 must be written with zeroes to allow normal TIMEKEEPER and RAM operation.

ADDRESS	DATA			FUNCTION	
	D, D, C	$D_5 D_4 D_3 D_1$	$D_2 D_1 D_0$		
7FF				YEAR	00-99
7FE	00	0		MONTH	01-12
7FD	00-			DATE	01-31
7FC	OFT	0 0 0		DAY	01-07
7FB	KS O -			HOUR	00-23
7FA	0			MINUTES	00-59
7F9	ST			SECONDS	00-59
7F8	WR	s		CONTROL	
KEY: ST = S	TOP BIT	R = R	EAD BIT	FT = FREQUENCY	TEST
W = W	RITE BIT	S = S	IGN BIT	KS = KICK START	

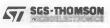
FIGURE 5. THE MK48T02/12 REGISTER MAP

Calibrating the Clock

The MK48T02/12 is driven by a quartz crystal controlled oscillator with a nominal frequency of 32768 Hz.The crystal is mounted in the tophat along with the battery. A typical MK48T02/12 is accurate within ± 1 minute per month at 25°C without calibration. The devices are tested not to exceed \pm 35 ppm (Parts Per Million) oscillator frequency error at 25°C, which comes to about ± 1.53 minutes per month. Of course the oscillation rate of any crystal changes with temperature. Figure 6 shows the frequency error or that can be expected at various temperatures.

Most clock chips compensate for crystal frequency and temperature shift error with cumbersome trim capacitors. The MK48T02/12 design, however, employs periodic counter correction. The calibration circuit adds or subtracts count from the oscillator divider circuit at the divide by 256 stage, as shown in Figure 7. The number of times pulses are blanked (subtracted, negative calibration) or split (added, positive calibration) depends upon the value loaded into the five bit Calibration byte found in the Control register. Adding count speeds the clock up, subtracting counts slows the clock down.

The Calibration byte occupies the five lower order bits in the Control register. The byte can be set to represent any value between 0 and 31 in binary form. The sixth bit is a sign bit; "1" indicates positive calibration, "0" indicates negative calibration. Calibration occurs within a 64 minute cycle. The first 62 minutes in the cycle may, once per minute, have one second either shortened or lengthened by 128 oscillator cycles, that is one tick of the divide by 256 stage. If a binary 1 is loaded into the register, only the first two minutes in the 64 minute cycle will be modified; if a binary 6 is loaded, the first 12 will be affected, and so on.



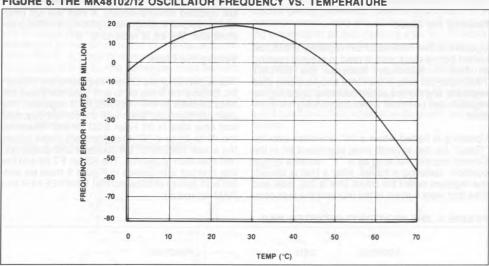
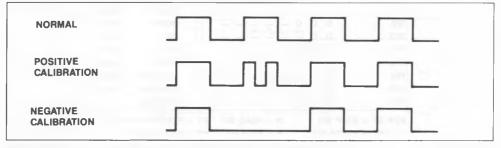


FIGURE 6. THE MK48T02/12 OSCILLATOR FREQUENCY VS. TEMPERATURE

FIGURE 7. ADJUSTING THE DIVIDE BY 256 PULSE TRAIN



Therefore, each calibration step has the effect of adding or subtracting 256 oscillator cycles for every 125,829,120 (32768 x 60 x 64) actual oscillator cycles, that is 2.034 ppm of adjustment per calibration step; giving the user a ± 63.07 ppm calibration range. Assuming that the oscillator is in fact running at exactly 32768 Hz, each of the 31 increments in the Calibration byte would represent 5.35 seconds per month.

Two methods are available for ascertaining how much calibration a given MK48T02/12 may require. The first involves simply setting the clock, letting it run for a month and comparing it to a known accurate reference (like WWV broadcasts). While that may seem crude, it allows the designer to give the end user the ability to calibrate his clock as his environment may require, even after the final product is packaged in a non-user serviceable enclosure. All the designer has to do is provide a simple utili ty that accesses the Calibration byte. The utility could even be menu driven and made foolproof.

The second approach is better suited to a manufacturing environment, and involves the use of some test equipment. When the Frequency Test (FT) bit, the seventh-most significant bit in the Day register, is set to a "1", and the oscillator is running at 32768 Hz, the LSB (DQ₀) of the Seconds register will toggle at a 512 Hz. Any deviation from 512 Hz indicates the degree and direction of oscillator frequency shift at the test temperature. For example, a reading of 512.00512 Hz would indicate a +10 ppm (1-(512/512.00512)) oscillator frequency error, requiring a -5 (0001012) to be loaded into the Calibration Byte for correction. Note that setting or changing the Calibration Byte does not affect the Frequency Test output frequency. The device must be selected and addresses must be stable at Address 7F9 when reading the 512 Hz on DQ₀.



The FT bit must be set using the same method used to set the clock, using the Write bit. The LSB of the Seconds register is monitored by holding the MK48T02/12 in an extended read of the Second register, without having the Read bit set. The FT bit MUST be reset to a "0" for normal clock operations to resume.

Stopping and Starting the Oscillator

The oscillator may be stopped at any time. If the device is going to spend a significant amount of time on the shelf, the oscillator can be turned off to minimize current drain from the battery. The "Stop" bit is the MSB of the Seconds register. Setting it to a "1" stops the oscillator. In order to make the oscillator as stingy with current as possible, the oscillator is designed to require an extra "kick" to begin oscillation again. The extra kick is provided by the Kick Start (KS) bit, the MSB of the Hours register. To start the oscillator, implement the following procedure.

- 1. Set the Write Bit to "1".
- 2. Reset the Stop Bit to "0".
- 3. Set the Kick Start Bit to "1".
- 4. Reset the Write Bit to "0".
- 5. Wait 2 seconds.
- 6. Set the Write Bit to "1".
- 7. Reset the Kick Start Bit to "0".
- 8. Set the Correct time and date.
- 9. Reset the Write Bit to "0".
- Note: Leaving the KS bit set will cause the Clock to draw excessive current and will shorten battery life.

DATA RETENTION MODE

With V_{CC} applied, the MK48T02/12 operates as a conventional BYTEWIDE static ram. However, V_{CC} is being constantly monitored. Should the supply voltage decay, the RAM will automatically powerfail deselect, write protecting itself when V_{CC} falls within the V_{PFD} (max), V_{PFD} (min) window. The MK48T02 has a V_{PFD} (max) · V_{PFD} (min) window of 4.75 volts to 4.5 volts, providing very high data security, particularly when all of the other system components are specified to 5.0 volts plus and minus 10%. The MK48T12 has a V_{PFD} (max) · V_{PFD} (min) window of 4.5 volts to 4.2 volts, allowing users constrained to a 10% power supply specification to use the device.

Note: A mid-write cycle power failure may corrupt data at the currently addressed location, but does not jeopardize the rest of the RAM's content. At voltages below V_{PFD} (min), the user can be assured the memory will be in a write protected state, provided the V_{CC} fall time does not exceed t_F. The

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MK48T02/12 may respond to transient noise spikes that reach into the deselect window if they should occur during the time the device is sampling V_{CC} . Therefore decoupling of power supply lines is recommended.

The power switching circuit connects external V_{CC} to the RAM and disconnects the battery when V_{CC} rises above V_{SO} . As V_{CC} rises the battery voltage is checked. If the voltage is too low, an internal Battery Not OK (BOK) flag will be set. The BOK flag can be checked after power up. If the BOK flag is set, the first write attempted will be blocked. The flag is automatically cleared after first write, and normal RAM operation resumes. Figure 8 illustrates how a BOK check routine could be structured.

Normal RAM operation can resume t_{REC} after V_{CC} exceeds V_{PFD} (Max). Caution should be taken to keep \overline{E} or W high as V_{CC} rises past V_{PFD} (Min) as some systems may perform inadvertent write cycles after V_{CC} rises but before normal system operation begins.



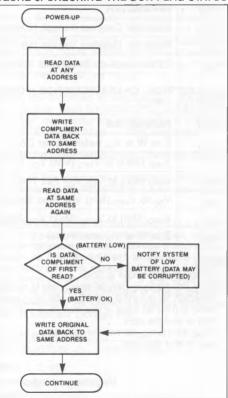
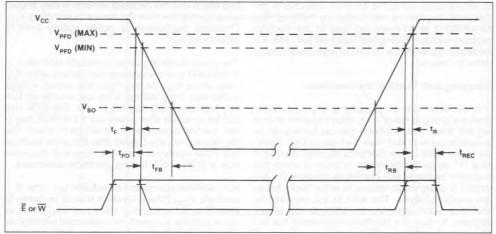


FIGURE 9. POWER-DOWN/POWER-UP TIMING



DC ELECTRICAL CHARACTERISTICS (POWER-DOWN/POWER-UP TRIP POINT VOLTAGES) (0 $^\circ C \leq T_A \leq +70 \,^\circ C)$

SYM	PARAMETER	MIN	ТҮР	MAX	UNITS	NOTES
V _{PFD}	Power-fail Deselect Voltage (MK48T02)	4.50	4.6	4.75	V	1
V _{PFD}	Power-fail Deselect Voltage (MK48T12)	4.20	4.3	4.50	V	1
V _{SO}	Battery Back-up Switchover Voltage		3		V	1

AC ELECTRICAL CHARACTERISTICS (POWER-DOWN/POWER-UP TIMING) (0 $^\circ C \le T_A \le +70\,^\circ C)$

SYM	PARAMETER	MIN	MAX	UNITS	NOTES
t _{PD}	E or W at V _{IH} before Power Down	0		ns	
t _F	V _{PFD} (Max) to V _{PFD} (Min) V _{CC} Fall Time	300		μs	2
t _{FB}	V _{PFD} (Min) to V _{SO} V _{CC} Fall Time	10		μS	3
t _{RB}	V _{SO} to V _{PFD} (Min) V _{CC} Rise Time	1		μS	
tR	V _{PFD} (Min) to V _{PFD} (Max) V _{CC} Rise Time	0		μS	
t _{REC}	E or W at V _{IH} after Power Up	2		ms	

NOTES:

- 1. All voltages referenced to GND.
- VPFD (Max) to VPFD (Min) fall times of less tF may result in deselection/write protection not occurring until 50
 µs after VCC passes VPFD (Min). VPFD (Max) to (Min) fall times of less than 10 µs may cause corruption of RAM data or stop the clock.
- VPFD (Min) to VSO fall times of less than tFB may cause corruption of RAM data or stop the clock.

CAUTION

Negative undershoots below -0.3 volts are not allowed on any pin while in Battery Back-up mode.





PREDICTING BACK-UP SYSTEM LIFE

The useful life of the battery in the MK48T02/12 is expected to ultimately come to an end for one of two reasons; either because it has been discharged while providing current to an external load; or because the effects of aging render the cell useless before it can actually be discharged. Fortunately, these two effects are virtually unrelated, allowing discharge, or Capacity Consumption and the effects of aging, or Storage Life to be treated as two independent but simultaneous mechanisms, the earlier of which defines Back-up System life.

The current drain that is responsible for Capacity Consumption can be reduced either by applying V_{CC} or turning off the oscillator. With the oscillator off, only the leakage currents required to maintain data in the RAM are flowing. With V_{CC} on, the battery is disconnected from the RAM. Because the leakage currents of the MK48T02/12 are so low, they can be neglected in practical Storage Life calculations. Therefore, application of V_{CC} or turning off the oscillator can extend the effective Back-up System IIfe.

Predicting Storage Life

Figure 10 illustrates how temperature affects Storage Life of the MK48T02/12 battery. As long as V_{CC} is applied or the oscillator is turned off, the life of the battery is controlled by temperature and is virtually unaffected by leakage currents drawn by the MK48T02/12.

Storage Life predictions presented in Figure 10 are extrapolated from temperature accelerated life-test data collected in over 100 million device hours of continuing bare cell and encapsulated cell battery testing by SGS-THOMSON. Obviously, temperature accelerated testing cannot identify non-temperature dependent failure mechanisms. However, in view of the fact that no random cell failures have been recorded in any of SGS-THOMSON's on going battery testing since it began in 1982, we believe the chance of such failure mechanisms surfacing is extremely small. For the purpose of this testing, a cell failure is defined as the inability of a cell stabilized at 25°C to produce a 2.0 volt closed-circuit voltage across a 250K ohm load resistance.

A Special Note: The summary presented in Figure 10 represents a conservative analysis of the data presently available. While SGS-THOMSON is most likely in possession of the largest collection of battery life data of this kind in the world, the results presented should not be considered absolute or final; they can be expected to change as yet more data becomes available. We believe that future read-points of life tests presently under way and improvements in the battery technology itself will result in a continuing improvement of these figures.

Two end of life curves are presented in Figure 10. They are labeled "Average" ($t_{50\%}$) and ($t_{1\%}$). These terms relate to the probability that a given number of failures will have accumulated by a particular point in time. If, for example, expected life at 70 °C is at issue, Figure 10 indicates that a particular MK48T02/12 has a 1% chance of having a battery failure 11 years into its life and a 50% chance of failure at the 20 year mark. Conversely, given a sample of devices, 1% of them can be expected to experience battery failure within 11 years; 50% of them can be expected to fail within 20 years.

The t_{106} figure represents the practical onset of wear out, and is therefore suitable for use in what would normally be thought of as a worst-case analysis. The $t_{50\%}$ figure represents "normal" or "average" life. It is, therefore, accurate to say that the average device will last " $t_{50\%}$ ".

Battery life is defined as beginning on the date of manufacture. Each MK48T02/12 is marked with a four digit manufacturing date code in the form YYWW (Example: 8625 = 1986, week 25).

Calculating Predicted Storage Life of the Battery

As Figure 10 indicates, the predicted Storage Life of the battery in the MK48T02/12 is a function of temperature.

Because the ambient temperature profile is dependent upon application controlled variables, only the user can estimate predicted Storage Life in a given design. As long as ambient temperature is held reasonably constant, expected Storage Life can be read directly from Figure 10. If the MK48T02/12 spends an appreciable amount of time at a variety of temperatures, the following equation should be used to estimate Storage Life.

Predicted Storage Life =

Т

Where TA_1 , TA_2 , TA_n = Time at Ambient Temperature 1, 2, etc.

$$T = Total Time = TA_1 + TA_2 + ... + TA_n$$

[(T

 SL_1 , SL_2 , SL_n = Predicted Storage Life at Temp 1, Temp 2, etc. (See Figure 10).



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Example Predicted Storage Life Calculation

A cash register/terminal operates in an environment where the MK48T02/12 is exposed to temperatures

of 30 °C (86 °F) or less for 4672 hrs/yr; temperatures greater than 25°C, but less than 40°C (104°F), for 3650 hrs/vr; and temperatures greater than 40 °C, but less than 70 °C (158 °F), for the remaining 438 hrs/yr.

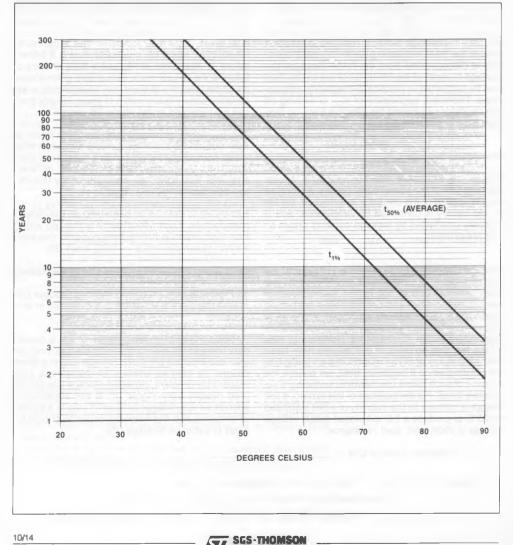
Reading predicted t_{196} values from Figure 10; SL₁ = 456 yrs., SL₂ = 175 yrs., SL₃ = 11.4 yrs.

Total Time (TT) = 8760 hrs./yr. TA₁ = 4672 hrs./yr. TA₂ = 3650 hrs./yr. TA₃ = 438 hrs./yr.

Predicted Typical Storage Life ≥

[(4672/8760)/456]+[(3650/8760)/175]+[(438/8760)/11.4] ≥ 126 yrs.

FIGURE 10. MK48T02/12 PREDICTED BATTERY STORAGE LIFE VS. TEMPERATURE



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Predicting Capacity Consumption Life

The MK48T02/12 internal cell has a minimum rated capacity of 35 mAh. The device places a nominal combined RAM and TIMEKEEPER load of 1.2 μ A on a typical internal 37 mAh lithium battery when the clock is running and the device is in Battery Back-up mode. At that rate, the MK48T02/12 will consume the cell's capacity in 29,166 hours, or about 3.3 years. But, as Figure 11 shows, Capacity Consumption can be spread over a much longer period of time.

Naturally, Back-up current varies with temperature. As Figure 12 indicates, the rate of Current Consumption by the MK48T02/12 with the clock running in Battery Back-up mode is a function of temperature. Because the ambient temperature profile is dependent upon application controlled variables, only the user can estimate consumption rates in a given design. As long as ambient temperature is held reasonably constant, expected Capacity Consumption life can be estimated by reading 0% V_{CC} Duty Cycle Capacity Consumption life directly from Figure 12, and dividing by the expected V_{CC} Duty Cycle (i.e. at 25°C with a 66% Duty Cycle, Capacity Consumption Life = 3.3(1-66) = 95 years).

If the MK48T02/12 spends an appreciable amount of time at a variety of temperatures, the same equation provided in the previous Storage Life section should be used to estimate Capacity Consumption life.

Example Consumption Life Calculation

Taking the same cash register/terminal used earlier, let's assume that the high and low temperature periods are the non-operating, Battery Back-up mode periods, and that the register is turned on 10 hours a day seven days per week. The two points of interest on the curves in Figure 12 will be the 25°C and the 70°C points.

Reading Capacity Life values from Figure 12; $CL_1 = 3.3$ yrs., $CL_2 = 3.55$ yrs.

Total Time (TT) = 8760 hrs./yr. $TA_1 = 4672$ hrs./yr. $TA_2 = 438$ hrs./yr.

[(4672/8760)/3.3]+[(438/8760)/3.55]

Estimating Back-up System Life

The procedure for estimating Back-up System Life is simple. Pick the lower of the two numbers. In the case calculated in the examples, that would be 5.69 years.

≥ 5.69 yrs.

The fact is, since either mechanism, Storage Life or Capacity Consumption, can end the system's life, the end is marked by whichever occurs first.



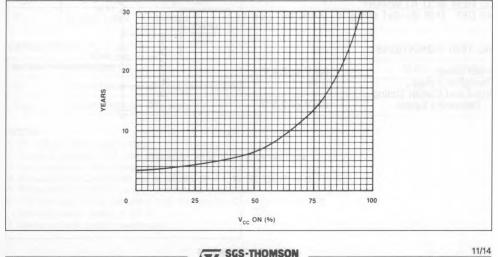




FIGURE 12. CURRENT CONSUMPTION LIFE OVER TEMPERATURE WITH 0% V_{CC} DUTY CYCLE

APPLICATION NOTE:

BINARY TO BCD, AND BCD TO BINARY CONVERSION

The MK48T02/12 presents and accepts TIMEKEEP-ER data in BCD format. Conversion to or from other formats can be executed in a single line of code, as the following example BASIC program demonstrates.

10 REM B	INARY TO BCD
20 DEF F	NA (X)=INT (X/10)*16+X-INT
	(/10)*10
30 REM B	CD TO BINARY
40 DEF F	NB (X)=INT (X/16)*10+(XAND15)

AC TEST CONDITIONS

Input Levels:	0.6 V to 2.4 V
Transition Times:	5 ns
Input and Output Timing	
Reference Levels	0.8 V or 2.2 V

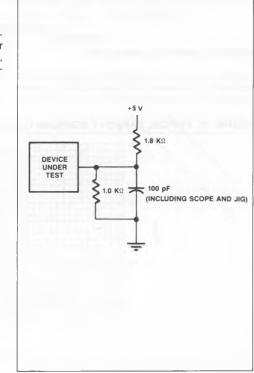


FIGURE 13. EQUIVALENT OUTPUT LOAD DIAGRAM



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ABSOLUTE MAXIMUM RATINGS*

Voltage On Any Pin Relative To GND0.3 V to +7.0 V
Ambient Operating (V_{CC} On) Temperature (T_A)
Ambient Storage (V _{CC} Off, Oscillator Off) Temperature
Total Device Power Dissipation 1 Watt
Output Current Per Pin
*Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device.

This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

RECOMMENDED DC OPERATING CONDITIONS

 $(0^{\circ}C \le T_A \le 70^{\circ}C)$

SYM	PARAMETER	MIN	MAX	UNITS	NOTES
V _{CC}	Supply Voltage (MK48T02)	4.75	5.50	V	1
V _{CC}	Supply Voltage (MK48T12)	4.50	5.50	V	1
GND	Supply Voltage	0	0	V	1
V _{iH}	Logic "1" Voltage All Inputs	2.2	V _{CC} + 0.3 V	V	1
VIL	Logic "0" Voltage All Inputs	-0.3	0.8	V	1,2

DC ELECTRICAL CHARACTERISTICS

 $(0^{\circ}C \le T_{A} \le +70^{\circ}C) (V_{CC} (Max) \ge V_{CC} \ge V_{CC} (Min))$

SYM	PARAMETER	MIN	MAX	UNITS	NOTES
I _{CC1}	Average V _{CC} Power Supply Current		80	mA	3
I _{CC2}	TTL Standby Current ($\overline{E} = V_{IH}$)		5	mA	4
I _{CC3}	CMOS Standby Current ($\overline{E} = V_{CC} - 0.2 V$)		3	mA	4
I _{IL}	Input Leakage Current (Any Input)	-1	+1	μA	5
IOL	Output Leakage Current	-5	+5	μA	5
V _{OH}	Output Logic "1" Voltage ($I_{OUT} = -1.0 \text{ mA}$)	2.4		V	Î
V _{OL}	Output Logic "0" Voltage (I _{OUT} = 2.1 mA)		0.4	V	

CAPACITANCE ($T_A = 25^{\circ}C$)

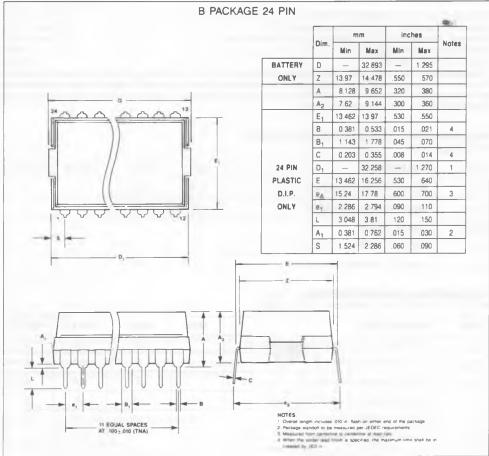
SYM	PARAMETER	MAX	NOTES
CI	Capacitance on all pins (except D/Q)	7 pF	6
C _{D/Q}	Capacitance on D/Q pins	10 pF	6,7

NOTES

- 1. All voltages referenced to GND.
- 2. Negative spikes of -1.0 volts allowed for up to 10 ns once per cycle.
- 3. ICC1 measured with outputs open.
- 4. Measured with Control Bits set as follows: R = 1; W, ST, KS, FT = 0.
- 5. Measured with $GND \le V_1 \le V_{CC}$ and outputs deselected.
- 6. Effective capacitance calculated from the equation C = $\frac{|\Delta t|}{V\Delta}$ with ΔV = 3 volts and power supply at 5.0 V.
- 7. Measured with outputs deselected.



PACKAGE DESCRIPTION



ORDERING INFORMATION

