# Secondary Side Average Current Mode Controller

#### FEATURES

- Practical Secondary Side Control of Isolated Power Supplies
- Provides a Self Regulating Bias Supply From a High Input Voltage Using an External N-Channel Depletion Mode FET
- Onboard Precision, Fixed Gain, Differential Current Sense Amplifier
- Wide Bandwidth Current Error
  Amplifier
- 5V Reference
- High Current, Programmable Gm Amplifier Optimized to Drive Opto-couplers

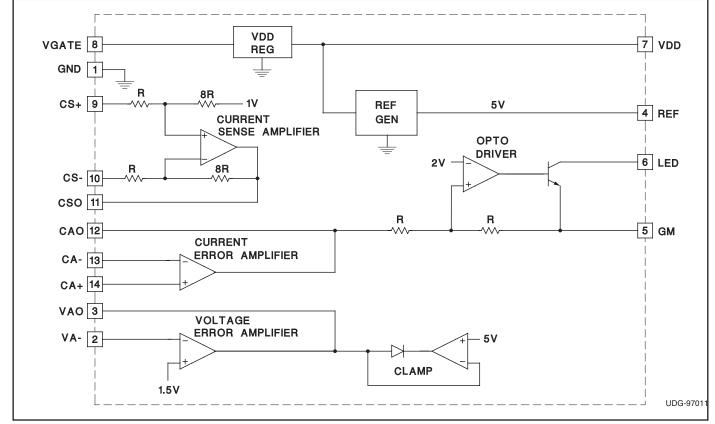
## DESCRIPTION

The UCC3839 provides the control functions for secondary side average current mode control in isolated power supplies. Start up, pulse width modulation and MOSFET drive must be accomplished independently on the primary side. Communication from secondary to primary side is anticipated through an opto-isolator.

Accordingly, the UCC3839 contains a fixed gain current sense amplifier, voltage and current error amplifiers, and a Gm type buffer/driver amplifier for the opto-isolator. Additional housekeeping functions include a precision 5V reference and a bias supply regulator.

Power for the UCC3839 can be generated by peak rectifying the voltage of the secondary winding of the isolation transformer. From this unregulated voltage, the UCC3839's bias supply regulator will generate its own 7.5V bias supply using an external, N-channel, depletion mode FET.

The UCC3839 can be configured for traditional average current mode control where the output of the voltage error amplifier commands the current error amplifier. It can also be configured for output voltage regulation with average current mode short circuit current limiting, employing two parallel control loops regulating the output voltage and output current independently.



## BLOCK DIAGRAM

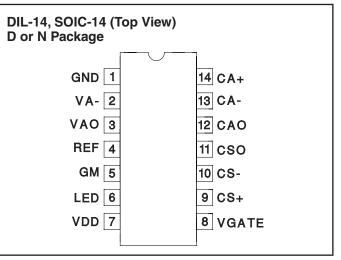
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#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage
Supply Current
(LED not connected) 2mA
(LED connected) 14mA
Analog Inputs0.3V to 15V
Power Dissipation at TA = 60°C
(LED not connected)
(LED connected)
Storage Temperature65°C to +150°C
Junction Temperature
Lead Temperature (Soldering, 10sec.)+300°C

Currents are positive into, negative out of the specified terminal. Consult Packaging Section of Databook for thermal limitations and considerations of package.

#### **CONNECTION DIAGRAMS**



**ELECTRICAL CHARACTERISTICS:** Unless otherwise specified, 0°C to 70°C for the UCC3839, -40°C to 85° for the UCC2839.  $V_{LINE} = 10V$ , RG = 400 $\Omega$ . T<sub>A</sub> = T<sub>J</sub>.

PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNITS
Current Error Amplifier					
VIO				10	mV
AVOL		60			dB
CMRR	VCM = 0.5V to 5.5V	60			dB
PSRR	VLINE = 10V to 20V	60			dB
CAO High	CA- = 1V, CA+ = 1.1V, ICAO = -100μA	4.8		7	V
ICAO	CA- = 1V, CA+ = 1.1V, CAO = 0.5V	-500		-250	μA
CAO Low	CA- = 1V, CA+ = 0.9V, ICAO = 500µA		0.2	0.4	V
GBW	F = 100kHz, TA = 25°C	3	5		MHz
Voltage Error Amplifier					-
VA-		1.46	1.5	1.525	V
AVOL		60			dB
PSRR	VLINE = 10V to 20V	60			dB
VAO High	IVAO = -100μA to 100μA	4.8	5	5.2	V
Ivao	VA- = 1.45V, VAO = 0.5V	-500		-250	μA
VAO Low	VA- = 1.55V, VAO = 0.5V, Ivao = 500µA		0.2	0.4	V
GBW	(Note 1)	3	5		MHz
Current Sense Amplifier					-
CSO Zero	CS+ = CS- = -0.3V to 5.5V, Icso = -100µA to 100µA	0.85	1	1.15	V
AV	CS+ = 0, CS- = 0mV to -200mV	7.75	8	8.2	V/V
Current Sense Amplifier (c	ont.)				•
Slew Rate	CS+ = 0, CS- = 0mV to -0.5V	2	4		V/µs
CSO	CS+ = -200mV, CS- = -700mV	4.75	5	5.2	V
LED Driver					
ILED	LED = 5.5V, CA- = 1V, CA+ = 1.1V, RG = 400		0	10	μA
	LED = 5.5V, CA- = 1V, CA+ = 0.9V, RG = 400	9	10	11	mA
Gm	LED = 5.5V, CAO = 1V to 3V, RG = 400	2.25	2.5	2.75	mS
Slew Rate	CAO = 2V to 2.5V, LED = 400Ω to 5.5V, RG = 400	1	4		V/µs

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PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Precision Reference					
REF	$T_J = 25^{\circ}C$	4.94	5	5.06	V
	IREF = 0mA to 1mA, VLINE = 10V to 20V	4.9		5.1	V
VA+/REF		0.298	0.3	0.302	V/V
VDD Regulator					
VDD	IDD = $0mA$ to $-15mA$ , VLINE = $10V$ to $40V$	7	7.5	8	V
IVDD	VLINE = 10V to 40V, CA- = 0V, CA+ = 1V, VA- = 2.9V, CS+ = CS- = 0, IREF = 0		1.3	2	mA

Note 1: Ensured by design. Not 100% tested in production.

#### **PIN DESCRIPTIONS**

CA-: Current Error Amplifier Negative Input.

**CAO:** Current Error Amplifier Output. Output source current is limited, and output sink current is guaranteed to be greater than the VAO output source current. Current loop compensation components are generally connected to CAO and CA–.

**CA+**: Current Error Amplifier Positive Input.

**CS-:** Current Sense Amplifier Negative Input.

**CSO:** Current Sense Amplifier Output. Internally set gain VOUT/VIN = 8 VIN = 0V results in CSO = 1V.

CS+: Current Sense Amplifier Positive Input.

**GM:** Gm (transconductance) Programming Pin. Resistor RGM =  $400\Omega$  to GND.

GND: Chip Ground.

**LED:** Output of LED Driver. Connect LED from VDD pin to LED.

**REF:** 5V Precision Reference Buffer Output. Minimum Decoupling Capacitance =  $0.01\mu$ F

**VA–:** Voltage Error Amplifier Negative Input. Voltage Error Amplifier is internally referenced to 1.5V

**VAO:** Voltage Error Amplifier Output. In a two loop average current mode control configuration, VAO is connected to CA+ and is the current command signal. VAO is internally clamped not to exceed 5V for short circuit control. In a single loop voltage mode control configuration with a parallel average short circuit current control loop, VAO is connected directly to CAO. Output source current is limited, and output sink current is guaranteed to be greater than the CAO output source current.

**VDD:** 7.5V Regulator output. Supply for most of the chip. Minimum Decoupling Capacitance =  $0.01\mu$ F

VGATE: External FET Gate Control Voltage.

#### **APPLICATION INFORMATION**

Fig. 1 shows a typical secondary side average current mode controller configuration using the UCC3839. In this configuration, output voltage is sensed and regulated by the voltage error amplifier. Its output, VAO provides the reference for the current error amplifier at the CA+ pin. VAO can be connected to CA+ directly or through a resistive divider depending on the particular application requirements.

Average current mode control needs accurate output current information which is provided by a low value current sense resistor. The voltage proportional to the converter's output current is sensed and amplified by the precision current sense amplifier of the chip. The onboard current sense amplifier has a gain of 8 and is intended for differential sensing of the shunt voltage with a common mode voltage range from 0V up to 5V. The output of the current sense amplifier, CSO is 1V for zero input which guarantees that the circuit can control currents down to 0A.

The CSO signal is fed to the CA– input of the current error amplifier through a resistor. The current error amplifier takes the VAO and CSO signals and generates the error signal for the pulse width modulator.

### **APPLICATION INFORMATION (cont.)**

Since the PWM function is located on the primary side of the power converter the CAO signal must be sent across the safety isolation boundary. The UCC3839 anticipates an opto-coupler to provide isolation between primary and secondary. Therefore, CAO drives a transconductance amplifier that controls LED current in an opto-isolator. During start up and when CAO exceeds 4V, the current in the LED drops to zero. Maximum LED current is obtained during normal operation as CAO reaches its lowest potential. Its value is determined by the programming resistor value from the GM pin to circuit GND.

An alternative secondary side controller configuration is introduced in Fig. 2. In this circuit, the voltage and current control loops of the UCC3839 are connected parallel. It can be achieved by connecting the VAO and CAO pins together. The error amplifier with the lower output voltage controls the current in the opto-coupler providing the feedback signal for the PWM section on the primary side. Voltage regulation is still maintained by the voltage error amplifier until a user programmable output current is reached. At this time CAO will take control over the Gm amplifier and the output current of the converter will be regulated while the output voltage falls below its nominal value. This current level is set at the CA+ input by a resistive divider from the 5V reference of the chip.

Since the chip is powered from a peak rectifier which maintains the bias supply for the UCC3839 even under short circuit conditions, both of these techniques can be used to eliminate the short circuit runaway problem in isolated power supplies using peak current mode control on the primary side.

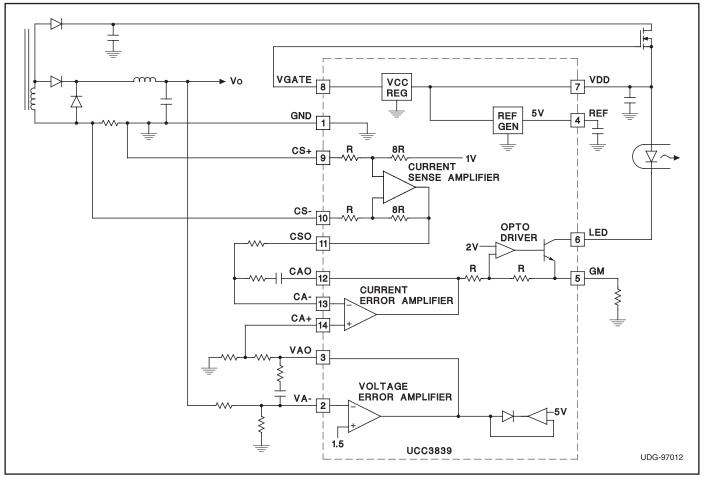


Figure 1. Secondary side average current mode controller.

## **APPLICATION INFORMATION (cont.)**

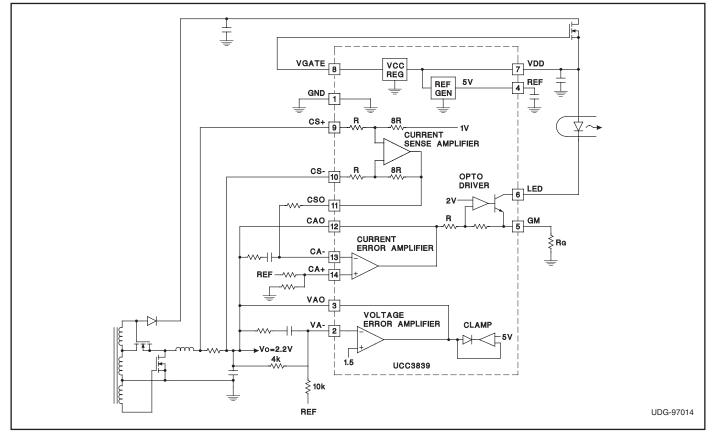


Figure 2. Voltage mode with average current short circuit limit.

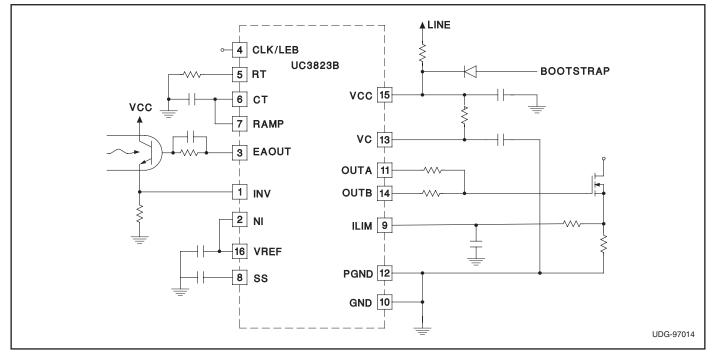


Figure 3. Typical primary side circuit for use with secondary side average current mode controller.



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Orderable Device	Status <sup>(1)</sup> F	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
UCC2839D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	Samples Not Available
UCC2839DG4	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	Samples Not Available
UCC3839D	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	Replaced by UC3849DW
UCC3839DG4	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	Samples Not Available

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<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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